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Design and Analysis of Discrete

Current Regulators for VSIs

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Netz Stabil

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Introduction

- This paper compares the quasi-continuous dq-decoupling techniques to various discrete modeling approaches for *Voltage Oriented Control* (VOC)
- To analyze the different techniques, this paper provides both *Frequency Response Functions* (FRFs) for *Command Tracking* (CT) and *Dynamic Stiffnes* (DS), and time domain tests
- Two *Voltage Source Inverter* (VSI) topologies have been used:
 1. Small-Scale Laboratory HVDC-MMC 2. Industrial Converter of a 3 MW Wind Turbine Generator

Compared dq-Decoupling Techniques



Discrete complex vector synchronous frame PI current Proposed discrete-time complex vector synchronous frame PI current regulator regulator obtained via direct modeling (Briz, et.al, 2000). With discrete DCCSFb and DID (red), MICC (blue), and MID (yellow).



Discrete-Time PI controller with DID and continuous DCCSFb Discrete State Block Diagram of the *RL*-plant of a VSI with approach in blue. MICC (blue) and MID (yellow) and full DSFb (red).

• Four different dq-decoupling techniques are presented and compared during dynamic events: 1. Discrete CCVC 2. Quasi-Continuous DCCSFb 3. Discrete DCCSFb w/ MID 4. Full DSFb

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Dynamic Analysis of the Control Techniques

---: $\tau_{est} = 3 \tau (L_{est} = 1.5L, R_{est} = 0.5R), \ \omega_e = 2\pi 500 \text{ rad/s}$ --: $\tau_{est} = \tau, \ \omega_e = 2\pi 500 \text{ rad/s}$ ---: $\tau_{est} = 3 \tau (L_{est} = 1.5L, R_{est} = 0.5R), \ \omega_e = 2\pi 50 \text{ rad/s}$ --: $\tau_{est} = \tau, \ \omega_e = 2\pi 50 \text{ rad/s}$

CT FRFs for 2kHz switching frequency, 500Hz bandwidth - (a) discrete CVC, (b) w/ continuous DCCSFb, (c) w/ discrete DCCSFb and MID.

Control technique with full DSFb; (a) control structure, (b) CT, (c) DS @ 2 kHz switching frequency, 500 Hz bandwidth.

---: $\tau_{est} = 3\tau (L_{est} = 1.5L, R_{est} = 0.5R), \ \omega_e = 2\pi 50 \text{ rad/s}$ ---: $\tau_{est} = \tau, \ \omega_e = 2\pi 50 \text{ rad/s}$

DS FRFs for 2 kHz switching frequency, 500Hz bandwidth - (a) discrete CVC, (b) w/ continuous DCCSFb, (c) w/ discrete DCCSFb and MID.

Response in *q*-axis current during a simultaneous step command on *q*- and *d*-axis current of the studied control topologies @ $\omega_e = 2\pi 500 \text{ rad/s}, f_{sw} = 2 \text{ kHz}$ and $\tau_{est} = 3\tau$ ($L_{est} = 1.5L, R_{est} = 0.5R$); (a) CVC, (b) discrete DCCSFb w/ MID, (c) discrete full DSFb w/ MID.

- Discrete approaches show superior command tracking and disturbance rejection properties
- The discrete techniques show different behavior regarding robustness
- The advantages of discrete approaches are more pronounced at lower switching frequencies

Delay Compensation and Analysis of Different Decoupling Techniques (State-Feedback-based)

Discrete plant model with one period input delay with delay compensation using a Discrete Luenberger Style Observer (DLSO).

Open-loop step response of various MID & DCCSFb techniques with parameter estimation error for a 3 MW two-level inverter with 1*Ts* input delay. $\tau_{est} = 1.5 \tau$ ($L_{est} = 1.2L, R_{est} = 0.8 R$) Step command: $v_{q,ref}(t) = 0$ V, $v_{d,ref}(t) = 50$ V, $\omega_e = 2\pi 50$ rad/s, w/ delayed parameters D_1 - D_4 .

- Discrete approaches provide enhanced decoupling properties compared to quasi-continuous (2)
- Technique (1) high robustness, but can get oscillatory at high output frequencies
- Technique (3) overall well-behaved. Medium sensitivity regarding delay and load estimation
- Technique (4) well-behaved. Low parameter sensitivity regarding load but very sensitive to delay

Conclusion

At low switching frequencies (depends on ratio of T_s and τ), the advantages of discrete modeling become very apparent, especially at high synchronous speed (method 2 is unstable for some o.p.)
The examined discrete decoupling techniques differ in robustness attributes

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